

Oxide-Relief and Zn-Diffusion 850-nm Vertical-Cavity Surface-Emitting Lasers With Extremely Low Energy-to-Data-Rate Ratios for 40 Gbit/s Operations

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Abstract—We demonstrate novel structures of a vertical-cavity surface-emitting laser (VCSEL) for high-speed (~ 40 Gbit/s) operation with ultralow power consumption performance. Downscaling the size of oxide aperture of VCSELs is one of the most effective ways to reduce the power consumption during high-speed operation. However, such miniaturized oxide apertures (~ 2 μm diameter) in VCSELs will result in a large differential resistance, optical single-mode output, and a small maximum output power (< 1 mW). These characteristics seriously limit the maximum electrical-to-optical (E–O) bandwidth and device reliability. By the use of the oxide-relief and Zn-diffusion techniques in our demonstrated 850-nm VCSELs, we can not only release the burden imposed on downscaling the current-confined aperture for high speed with low-power consumption performance, but can also manipulate the number of optical modes inside the cavity to maximize the E–O bandwidth and product of bit-rate transmission distance in an OM4 fiber. State-of-the-art dynamic performances at both room temperature and 85 °C operations can be achieved by the use of our device. These include extremely high D-factors (~ 13.5 GHz/mA^{1/2}), as well as record-low energy-to-data ratios (EDR: 140 fJ/bit) at 34 Gbit/s operation, and error-free transmission over a 0.8-km OM4 multimode fiber with a record-low energy-to-data distance ratio (EDDR: 175.5 fJ/bit.km) at 25 Gbit/s.

Index Terms—Semiconductor lasers, vertical-cavity surface-emitting lasers (VCSELs).

I. INTRODUCTION

HIGH-SPEED, high-efficiency, and low power consumption vertical-cavity surface-emitting lasers (VCSELs) that operate at a wavelength of 850 nm or around 1000 nm have lately attracted a lot of attention due to their suitability for applications in optical interconnects (OI) [1]–[12]. Compared with

the reported edge-emitting distributed feed-back (DFB) laser with its state-of-the-art dynamic performance at a wavelength of 1.3 μm [13]–[15] which is used for an OI application, the power consumption of 850 or ~ 1000 nm VCSELs for > 25 Gbit/s operation is less than 1/10 [16]. There are two important parameters necessary for evaluating power consumption during high-speed operation of VCSELs.

One is the modulation current efficiency (D-factor) and the other is the threshold current I_{th} [1]–[12]. In order to further reduce the required power consumption I_{th} and increase the D-factor of VCSELs, there are two major trends under development. The one involves using the highly strain multiple quantum wells (MQWs) to increase the differential gain in the active layers and improve the D-factor performance [5], [6], [8], [10]–[12], [17]. Another approach is to greatly downscale the size of the oxide aperture to reduce I_{th} [3], [4]. Recently, by downscaling the diameter of the oxide aperture to around 2 μm , record-low energy-to-data ratio (EDR: 83 fJ/bit) of 850-nm single-mode VCSELs with excellent transmission performance under 17 Gbit/s and at room temperature (RT) operation has been demonstrated [3]. However, compared with the reported high-performance 850-nm VCSEL with a larger oxide aperture (~ 6 μm), a significant degradation in the maximum 3-dB electrical-to-optical (E–O) bandwidth is observed (23 Gbit/s [2] versus ~ 13 Gbit/s [3]). Such degradation in speed performance might be attributable to the tremendous increase in the differential resistance of a VCSEL (~ 570 Ω [3]) with such a small oxide aperture (~ 2 μm), the single-mode output-induced spatial hole burning effect [18]–[20], and the relaxation oscillation (RO) frequency-limited E–O bandwidth [19]–[21]. In addition, the significant stress of (double) oxide layers [1], [2], [8] on the active regions should be regarded as a more serious issue impacting the reliability of these VCSELs when their current-confined oxide apertures [22] are further downscaled. In this paper, we demonstrate a novel high-speed VCSEL that operates at the 850-nm wavelength, incorporating oxide-relief [12] and Zn-diffusion [11], [19]–[21] structures to overcome the aforementioned limitations of VCSELs with miniaturized oxide apertures. By selective removing the single-oxide layer in our VCSEL structure, the reduction in the parasitic capacitance, the improvements in the D-factor, and maximum modulation speed have been demonstrated [12]. Furthermore, in order to further reduce the device resistance, manipulate the number of optical modes inside the VCSEL cavity,

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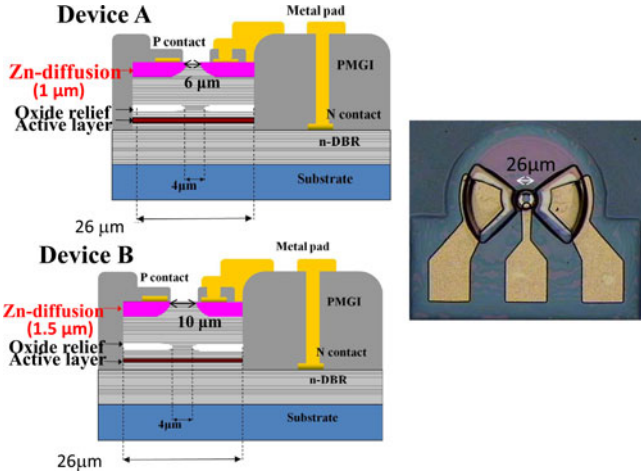


Fig. 1. (a) Conceptual cross-sectional view of devices A and B, and (b) top view of the demonstrated device.

and maximize both the E–O bandwidth and product of bit-rate transmission distance in the OM4 fiber, the Zn-diffusion techniques are applied to the top DBR mirror. The VCSEL with an $\sim 4\text{-}\mu\text{m}$ diameter current-confined (oxide) aperture and optimized Zn-diffusion depth demonstrates state-of-the-art performance. It includes extremely large D-factor ($13.2\text{ GHz/mA}^{1/2}$), record-low energy-to-data ratios (EDR: 140 fJ/bit) at 34 Gbit/s operation [1]–[12], and 25 Gbit/s error-free transmission over a 0.8-km OM4 multi-mode fiber with a record-low energy-to-data distance ratio (EDDR: 175.5 fJ/bit.km) [23], [24].

II. DEVICE STRUCTURE AND FABRICATION

Fig. 1(a) and (b) shows the conceptual cross-sectional and top views of the demonstrated device, respectively. As can be seen in Fig. 1(b), the fabricated device has a $26\text{-}\mu\text{m}$ diameter active mesa, which is integrated with co-planar waveguide (CPW) pads for on-wafer high-speed measurement. The epilayer structure grown on a semi-insulating GaAs substrate is composed of three $\text{In}_{0.15}\text{Al}_{0.1}\text{Ga}_{0.75}\text{As}/\text{Al}_{0.3}\text{Ga}_{0.7}\text{As}$ MQWs sandwiched between 30-pair n-type and 20-pair p-type $\text{Al}_{0.9}\text{Ga}_{0.1}\text{As}/\text{Al}_{0.12}\text{Ga}_{0.88}\text{As}$ distributed-Bragg-reflector (DBR) layers with an $\text{Al}_{0.98}\text{Ga}_{0.02}\text{As}$ layer above the MQWs for oxidation [11]. Compared with that reported for $\text{In}_{0.1}\text{Ga}_{0.9}\text{As}/\text{Al}_{0.37}\text{Ga}_{0.63}\text{As}$ strained 850-nm VCSELs [17], we can achieve a higher indium mole fraction (0.15 versus 0.1) with a thicker well width (50 \AA versus 40 \AA) due to the incorporation of Al, which enlarges the bandgap but has no significant influence on the lattice constant (strain) of the well layer. This means that our strained VCSEL structure has a higher indium mole fraction and a wider well width than those of InGaAs-based strained 850-nm VCSELs [17] which may lead to a more significant improvement in static/dynamic performance and uniformity of wafer, respectively. An oxidation technique is used to define a circular current-confined area $4\text{ }\mu\text{m}$ in diameter. Here, the single-oxide layer is adopted in our structure simply to realize the oxide-relief structure. By the use of a chemical solution, which offers excellent etching selectivity between the $\text{Al}_x\text{O}_{1-x}$

layer and active (Al(In)GaAs) layers, we can thus get a very high yield ($>90\%$) with the fabricated device. Due to the fact that the dielectric constant of air is just around one-half that of the $\text{Al}_x\text{O}_{1-x}$ layer, there is a further reduction in the parasitic capacitance, as confirmed by microwave scattering (S) parameter measurement and device modeling techniques carried out on the oxide-relief and control VCSELs. The measurement results clearly indicate that the oxide-relief structure can effectively reduce the parasitic capacitance of the VCSEL. Details about the device modeling process for our VCSELs can be referred to our previous works [20]. Fig. 1(a) shows the Zn-diffusion structure with two different geometric sizes (devices A and B) realized in the top DBR of our oxide-relief VCSEL, which can further reduce the device resistance due to the disordering of the DBR structure after Zn-diffusion [11], [19]–[21]. Furthermore, we can also control the depth of Zn-diffusion to manipulate the number of optical modes inside our cavity [20], which has a significant influence on the transmission experiment as will be discussed later.

The diameter and depth of the Zn-diffusion aperture in device A is $6\text{ }\mu\text{m}$ and $\sim 1\text{ }\mu\text{m}$, respectively. The $6\text{-}\mu\text{m}$ Zn-diffusion aperture with a $>1.5\text{ }\mu\text{m}$ diffusion depth will ensure stable single-mode output performance in 850-nm VCSELs [11], [19], [20]. During high-speed data transmission, the single-mode output might reduce the mode dispersion in the fiber and benefit the product of bit-rate transmission distance [23], [24]. However, the pure single-mode performance in the VCSEL usually induces a serious spatial hole burning effect, which will result in a serious low-frequency roll-off and limit the maximum E–O bandwidth of the VCSEL [19]–[21]. In device A, we thus adopt a Zn-diffusion depth of less than $1\text{ }\mu\text{m}$, to avoid the pure single-mode performance and balance the tradeoff between the maximum modulation speed and fiber transmission distance. On the other hand, in order to further reduce the differential resistance of the VCSEL and improve the RC-limited bandwidth, a larger Zn-diffusion depth (time) is desired. In order to avoid the pure single-mode output performance, in device B, we adopt a Zn-diffusion aperture with a depth of $1.5\text{ }\mu\text{m}$ but a diameter larger than $6\text{ }\mu\text{m}$ ($\sim 10\text{ }\mu\text{m}$). We can thus expect a smaller differential resistance for device B than device A.

III. MEASUREMENT RESULTS

Fig. 2 shows the light output versus current (L – I) and the bias current versus voltage (I – V) characteristics of devices A and B. As can be seen, both devices exhibit the same low threshold current ($\sim 0.4\text{ mA}$). Such a result clearly indicates that our Zn-diffusion process does not seriously degrade the threshold current performance [11], [19], [20]. Furthermore, we can clearly see that device B has a lower differential resistance than that of device A due to its larger Zn-diffusion depth, as expected. This result leads to an improvement in the output power and differential quantum efficiency under a high bias current ($I > \sim 4\text{ mA}$). Overall, the differential resistance achieved by these two devices (A and B) can be as low as $\sim 140\text{ }\Omega$ (under 6 mA bias current). Such a value is smaller than the values

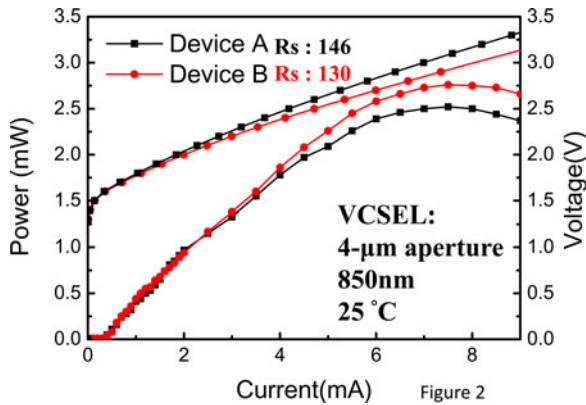


Fig. 2. Characteristics of the optical output power L and voltage V versus biasing current I of devices A and B.

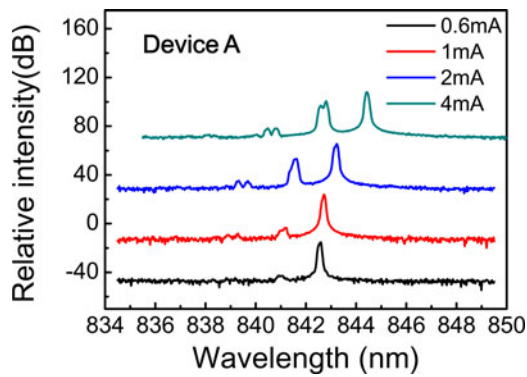


Fig. 3. Measured optical spectra of device A under different bias currents.

typically reported for high-speed VCSELs for >25 Gbit/s operation ($>200 \Omega$) [1]–[5], [8].

During dc measurement, we have tested over 30 VCSELs each for device structures A and B, which were fabricated on two different chips. The number of total devices per chip was around 100.

Around 80% of the L – I and I – V curves measured for these devices exhibit similar characteristics and trends to those illustrated in Fig. 2. This evidence clearly indicates the uniformity and repeatability of the Zn-diffusion and oxide-relief processes, which also supports our conclusion regarding the device measurement results, as will be discussed later. Figs. 3 and 4 show the optical spectra for devices A and B measured under different bias currents. As can be seen, under a very small bias current ($I < 0.6$ mA), both devices exhibit (near) single-mode behavior. On the other hand, when the bias current is further increased, both devices exhibit multimode performance, with device B having a larger number of optical modes due to its larger diameter of Zn-diffusion aperture ($10 \mu\text{m}$ versus $6 \mu\text{m}$). The different number of optical modes between these two devices would have a determinant influence on their performance in 25 Gbit/s data transmission, as will be discussed later in Figs. 14 and 15.

The high-speed E–O performance of the fabricated devices was measured by a lightwave component analyzer (LCA), which was composed of a network analyzer (Anritsu 37397 C) and a calibrated 25-GHz photoreceiver module (New focus 1481-S).

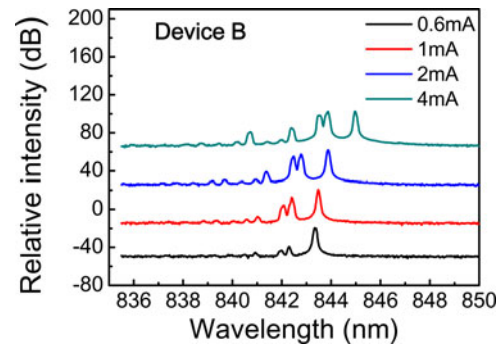


Fig. 4. Measured optical spectra of device B under different bias currents.

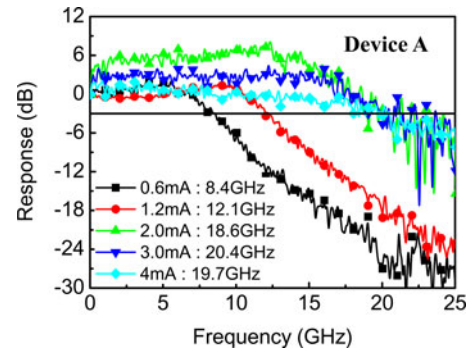


Fig. 5. Measured E–O frequency responses of device A under different bias currents and RT operation.

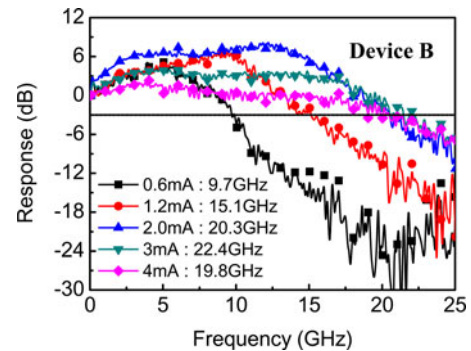


Fig. 6. Measured E–O frequency responses of device B under different bias currents and RT operation.

Figs. 5 and 6 show the E–O responses of devices A and B, respectively, under different bias currents at RT. As can be seen, both these two devices can achieve ~ 21 GHz 3-dB bandwidth with bias currents as low as ~ 3 mA. Such a value is just around one-half of the required bias current reported for a 40 Gbit/s 850-nm VCSEL with a similar O–E bandwidth performance (~ 22 GHz at 8 mA) [1], [2]. In addition, the maximum 3-dB E–O bandwidth of device B is slightly higher than that of device A (22 GHz versus 21 GHz). This can be attributed to the lower differential resistance of device B, as shown in Fig. 1, and a higher RC-limited bandwidth can thus be expected. Fig. 7(a) and (b) shows the relaxation oscillation (RO) frequency (f_R) (based on the shown E–O response in Fig. 4) versus $\sqrt{I - I_{th}}$ of these two devices, where I is the bias current and I_{th} is the

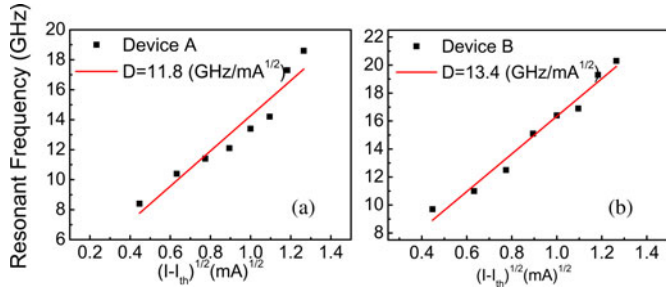


Fig. 7. Measured f_R versus $(I - I_{th})^{1/2}$ and extracted D-factors of (a) device A and (b) device B.

threshold current. The slope D of these fitting lines can be used to evaluate how fast f_R increases with the bias current. A slightly larger D-factor is extracted for device B than that for device A. Its value can be as large as $13.4 \text{ GHz/mA}^{1/2}$, which is even higher than that of the reported high-performance 850-nm VCSELs [3] which have a smaller diameter of oxide aperture ($4 \mu\text{m}$ versus $2 \mu\text{m}$). However, our measurement result here is somewhat in conflict with that obtained in previous work [19], [20], which stated that the VCSEL with a (near) single-mode (smaller number of optical modes) performance would have a larger D-factor than that of the highly multimode VCSEL, due to the minimization of the spreading of injected carriers into different optical modes during operation. This contradiction can be explained as follows. As reported in our previous work [20], the near single-mode VCSEL has a significant larger D-factor than that of the multi-mode control only under a small bias current regime ($< 2 \text{ mA}$). When the bias current is further increased, the E–O bandwidth of the single-mode VCSEL becomes saturated and limited by f_R [21], but the bandwidth of the multimode VCSEL can still increase with the increase in the bias current [20] until eventually exhibiting a much larger maximum 3-dB bandwidth (17 GHz versus 12.5 GHz) than that of the single-mode device. This would result in a larger D-factor for the multimode device under the high bias regime. This nonlinear relation between the bias current and E–O bandwidth suggests that we may not precisely predict the large-signal (large current swing) modulation characteristics of VCSELs just by the use of the extracted single D-factor value.

Although device A has a smaller number of optical modes and higher carrier injection efficiency than device B, its near single-mode characteristic demonstrates a more serious f_R -limited 3-dB bandwidth and degradation in the maximum 3-dB E–O bandwidth as well as the values of the D-factor. We thus measure a slightly larger D-factor for device B than for device A.

Figs. 8–11 show the measured $-\log$ [bit error rate (BER)] versus V_{pp} of these two devices under different data rates (12.5, 25, $\sim 40 \text{ Gbit/s}$) and different bias current operations at RT. Their insets show the corresponding error-free eye patterns under the operation conditions with the lowest bias current (power consumption).

For eye-pattern measurement, we adopted a 25-GHz photoreceiver module (New focus 1481-S) connected with a wideband (100 kHz to 65 GHz) radio-frequency (RF) amplifier (Centel-

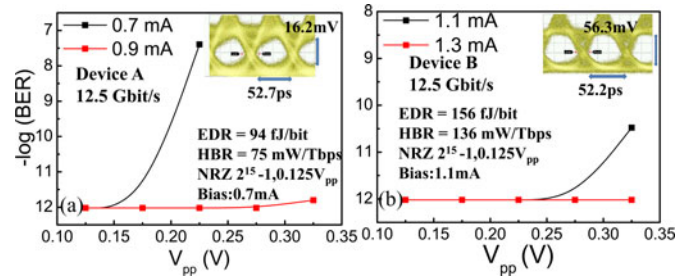


Fig. 8. Measured values of $-\log(\text{BER})$ at 12.5 Gbit/s versus RF peak-to-peak driving voltage (V_{pp}) of (a) device A and (b) device B under different bias currents.

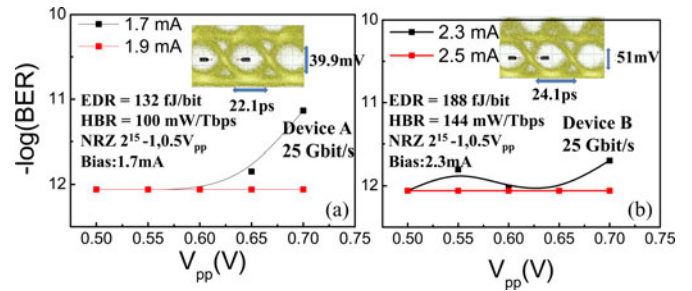


Fig. 9. Measured values of $-\log(\text{BER})$ at 25 Gbit/s versus RF peak-to-peak driving voltage (V_{pp}) of (a) device A and (b) device B under different bias currents.

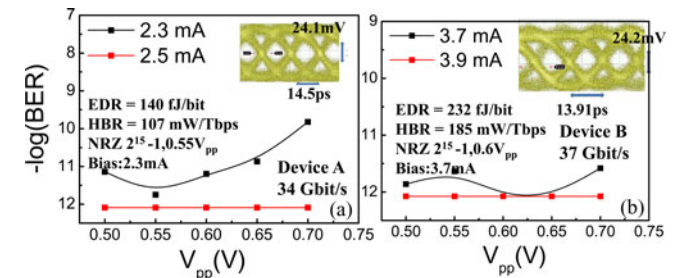


Fig. 10. Measured values of $-\log(\text{BER})$ at around 35 Gbit/s versus RF peak-to-peak driving voltage (V_{pp}) of (a) device A and (b) device B under different bias currents.

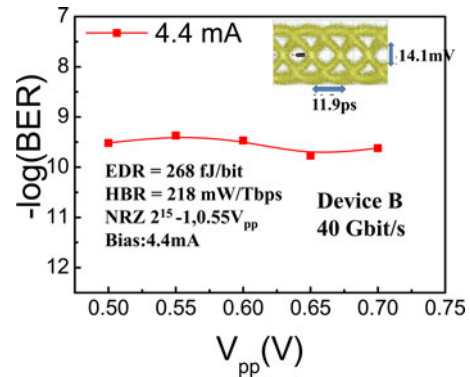


Fig. 11. Measured values of $-\log(\text{BER})$ at 40 Gbit/s versus RF peak-to-peak driving voltage (V_{pp}) of device B under different bias currents.

lax, UA0L65VM) to serve as our 40 Gbit/s receiver. Here, we use three parameters—energy-to-data rate (EDR) ratio (fJ/bit),

heat-to-bit rate (HBR) ratio (mW/Tbs), and energy-to-data distance (EDDR) ratio (fJ/(bit·km)) [23], [24]—to evaluate the eye-pattern performance of our devices. For a detailed definition of these three parameters, refer to [23]. A duplicate of these three definitions follows

$$\text{EDR} = \frac{P_e}{\text{BR}} \quad (1)$$

$$\text{HBR} = \frac{P_{\text{diss}}}{\text{BR}} \quad (2)$$

$$\text{EDR} = \frac{P_e}{(\text{BR} \times d)} \quad (3)$$

where $P_e = V \times I$ is the total consumed dc electrical power, V and I are the VCSEL's bias voltage and current during high-speed modulation, respectively, $P_{\text{diss}} = P_e - P_{\text{optical}}$ is the dissipated power, P_{optical} is the output optical power, BR is the bit rate, and d is the transmission distance in the optical fiber.

As can be seen, there is a significant increase in the values of BER for both devices with the driving voltage under the low-bias current regime (~ 1 mA), as shown in Figs. 8 and 9. This phenomenon mainly arises because, under such a low-bias current, which is near the threshold current, the operation point of the device is close to the highly nonlinear regime of the L - I curve. The larger driving voltage (current swing) leads to more serious distortion of the eye patterns and increases the BER values. On the other hand, when there is a further increase in the bias current, the operation point is closer to the linear regime (center) of the L - I curve. The influence of the magnitude of the driving voltage on the waveform quality of the eye pattern and the BER values can thus be minimized.

Furthermore, compared with that of device A, due to the smaller differential resistance and higher maximum E–O bandwidth, device B can achieve a higher maximum error-free data rate (37 Gbit/s versus 34 Gbit/s), as shown in Fig. 10(a) and (b). Nevertheless, device A exhibits smaller values of EDR and HBR compared with those of device B. Under RT operation, device A can achieve error-free 34 Gbit/s eye pattern under a bias current as low as 2.3 mA with an extremely small 0.55 V peak-to-peak driving voltage (V_{pp}).

The superior power consumption performance of device A compared to device B can mainly be attributed to the smaller number of optical modes inside the VCSEL cavity of device A, which would minimize the spreading of injected carriers into different optical modes, enhance the large-signal modulation efficiency, and possibly minimize the mode-selective loss, which would lead to degradation in the quality of the measured eye pattern [18], [20], [25]. Although device B has a slightly larger D-factor (modulation current efficiency) compared with device A, it cannot be used to accurately evaluate the large-signal eye-pattern performance of both devices, as discussed in Fig. 7.

In addition, as shown in Fig. 11, by further boosting the bias current of device B up to 4.4 mA, a near error-free 40 Gbit/s operation can be achieved. Such 40 Gbit/s BER values are limited by the intersymbol interference (ISI) noise from our VCSEL. When the injected data rate reaches 40 Gbit/s, there is a slight degradation in the measured signal-to-noise ratio of the eye pat-

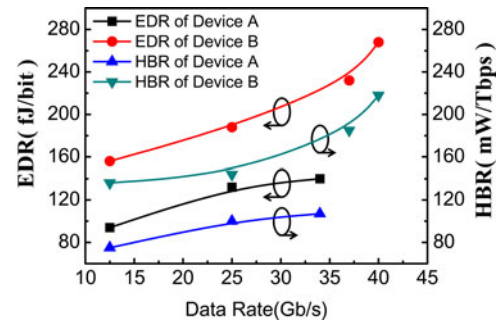


Fig. 12. Measured values of EDR and HBR of devices A and B versus different data rates

tern due to the limited E–O bandwidth of device B. As can be seen, the maximum bandwidth of device B is around 22.4 GHz, which is marginal for 40 Gbit/s error-free operations [1]. It should be possible to further push our 40 Gbit/s BER values to the error-free regime ($\text{BER} < 1 \times 10^{-12}$) by using a photoreceiver module with an improved sensitivity at 40 Gbit/s. At the receiver end, as adopted here, there is no transimpedance amplifier (TIA) integrated with our photodiode for 40 Gbit/s eye-pattern measurement (New focus 1481-S).

Fig. 12 shows all detailed values of EDR and HBR for these two devices under different data rates of operation. We can clearly see that the required values of EDR and HBR gradually increase with the data rate and record low values are achieved (EDR: 140 fJ/bit, HBR: 107 mW/Tbps) at 34 Gbit/s by device A among all the reported high-speed VCSELs with an operation speed close to 40 Gbit/s [1], [2], [4]–[6], [8]. Table I shows the benchmark for the reported high-speed VCSELs with their maximum data rate, EDR, and HBR. Our achieved values of EDR/HBR are close to the results reported in [3] but at a significant higher data rate (34 Gbit/s versus 25 Gbit/s).

High-speed VCSEL performance under 85 °C operation is also an important issue for practical applications of VCSELs with miniaturized sizes of oxide aperture [3], [24]. Fig. 13(a) and (b) shows the measured error-free eye diagram for devices A and B at their maximum modulation speed under 85 °C operation. Compared with the measurement results under RT operation, we can clearly observe that the required bias current increases slightly (2.3 to 2.9 mA for device A) and the maximum modulation speed degrades from 34 (37) to 32 (34) Gbit/s for device A (B).

The high-speed performance under high-temperature operation of our device is superior to that reported for multioxide 850-nm VCSELs with a miniaturized oxide aperture of around $\sim 2 \mu\text{m}$ [3], [24]. This is because the Zn-diffusion technique allows the differential resistance of the VCSEL to be significantly reduced, which implies less device heating under high-temperature operation.

Due to the increase in the area of modern data centers, it is important to increase the maximum transmission distance to around 1 km of OI [23], [24]. However, the bandwidth–distance product of standard OM4 multimode fibers is limited to around 4.5 GHz·km transmission distance [26]. Figs. 14 and 15 show the $-\log(\text{BER})$ measured at 25 Gbit/s operation versus the

TABLE I
BENCHMARK FOR HIGH-SPEED AND GREEN VCSELS

	TUB/VIS ³ (850nm)	Furukawa ¹⁰ (1060nm)	Chalmers ¹ (850nm)	UCSB ⁸ (980nm)	This work* (850nm)	TUB/VIS ⁴ (980nm)	NEC ⁵ (1100nm)
Max Bit rate(Gb/s)	25	10	40	35	34 (40)	35	30
Max EDR (fj/bit)	117	180	600	357	140 (268)	287	302
Max HBR (mW/Tbps)	99	140	452	286	107 (218)	233	265

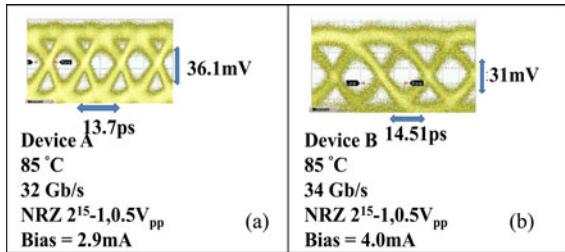


Fig. 13. Measured eye patterns and corresponding operation conditions at 85 °C of (a) device A and (b) device B at their maximum operation speed.

optical power (receiver end) and injected bias current (transmitter side) of devices A and B, respectively. The multimode fiber we adopted is the standard OM4 fiber. As can be seen, device A exhibits a much longer error-free transmission distance (0.8 km versus 0.1 km) than device B at the same data rate (25 Gbit/s). The EDDR achieved by device A is as low as 175.5 fJ/bit-km. To the best of the authors' knowledge, this number is the record reported for all VCSELs [24] for 25 Gbit/s data transmission through multimode fibers. The superior transmission performance of device A compared to device B can be attributed to the fact that device A has a smaller number of optical modes, which minimize the serious mode dispersion in the multimode fiber for high-speed data transmission [23], [24]. In addition, as shown in Fig. 14, device A exhibits a distinct behavior in the measured trace of $-\log(\text{BER})$ versus bias current. As can be seen, there is a wide bias current range (1.8–2.7 mA) for error-free transmission performance and a further increase in the bias current (>2.7 mA) would lead to degradation in the BER performance. This result can be attributed to the fact that the increase in bias current induces a more significant multimode phenomenon in the optical spectra, as shown in Figs. 3 and 4, which can seriously degrade transmission performance due to the mode dispersion. Such a result clearly indicates that the output optical spectra of high-speed VCSELs play an important role in 25 Gbit/s transmissions through OM4 multimode fibers. In addition, in order to investigate the 25 Gbit/s transmission performance at 85 °C, we also measure the optical spectra of device A when the ambient temperature rises to 85 °C. The optical spectra measured at RT and 85 °C are very similar in terms of number of modes and side-mode suppression ratio (SMSR), except for the red-shift at the center wavelength. Both measurement results, for the optical spectra and dynamic speed performance at high temperature, as shown in Fig. 13, confirm

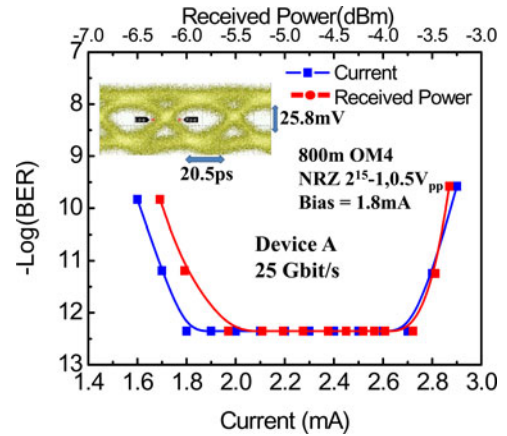


Fig. 14. Measured $-\log(\text{BER})$ at 25 Gbit/s through 0.8-km OM4 fiber transmission versus received optical power and injected bias current of device A.

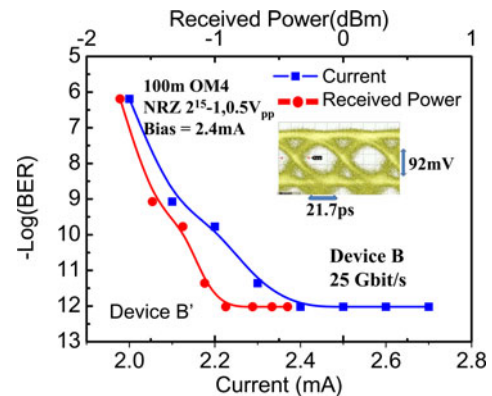


Fig. 15. Measured $-\log(\text{BER})$ at 25 Gbit/s through 0.1-km OM4 fiber transmission versus received optical power and injected bias current of device B.

the feasibility of using device A for 25 Gbit/s data transmission at 85 °C.

IV. CONCLUSION

In conclusion, we demonstrate a novel VCSEL structure for high-speed, ultralow power consumption, and short-reach (~ 1 km) data transmission through standard OM4 multimode fibers. By using the Zn-diffusion and oxide-relief structures in our demonstrated VCSEL, we can greatly reduce the large differential resistance of VCSELs with a miniaturized oxide aperture, further enhance the RC-limited bandwidth of the

device, and manipulate the optical mode inside cavity to improve the transmission performance. State-of-the-art dynamic performances, which include record-low EDR: 140 fJ/bit at data rate as high as 34 Gbit/s and record-low EDDR: 175.5 fJ/bit.km at 25 Gbit/s transmission over a standard OM4 fiber, can be achieved by the use of our demonstrated device with optimized sizes of Zn-diffusion and oxide-relief aperture.

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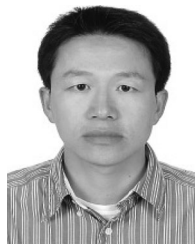
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